

CHE302 LECTURE IX FREQUENCY RESPONSES

Professor Dae Ryook Yang

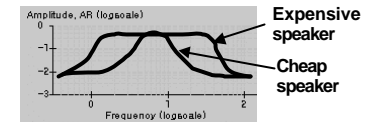
Fall 2001

Dept. of Chemical and Biological Engineering
Korea University

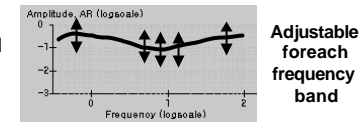
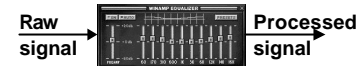
BENEFITS OF FREQUENCY RESPONSE

- Frequency responses are the informative representations of dynamic systems

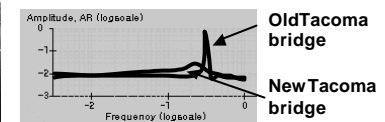
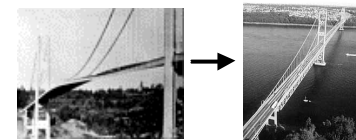
- Audio Speaker



- Equalizer



- Structure

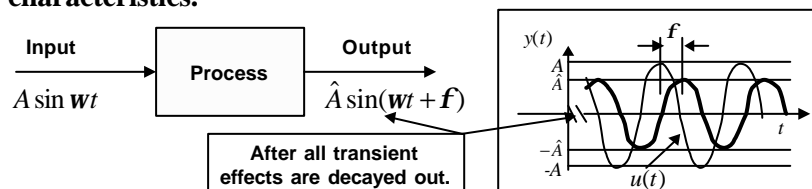


Old Tacoma bridge New Tacoma bridge

DEFINITION OF FREQUENCY RESPONSE

- For linear system

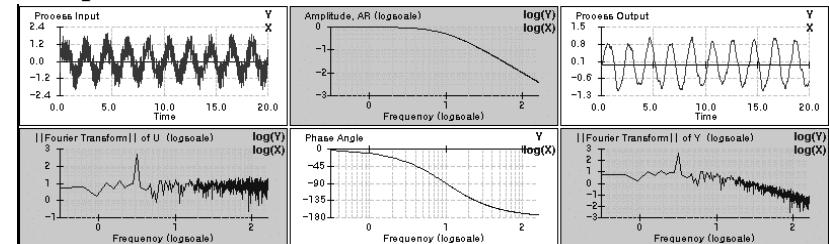
- “The ultimate output response of a process for a sinusoidal input at a frequency will show amplitude change and phase shift at the same frequency depending on the process characteristics.”



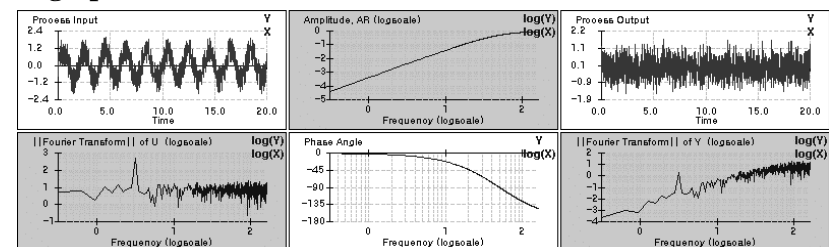
After all transient effects are decayed out.

- Amplitude ratio (AR): attenuation of amplitude, \hat{A} / A
- Phase angle (f): phase shift compared to input
- These two quantities are the function of frequency.

- Low-pass filter



- High-pass filter

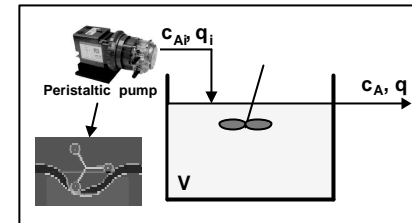


- In signal processing field, transfer functions are called “filters”.

- Any linear dynamical system is completely defined by its frequency response.
 - The AR and phase angle define the system completely.
 - Bode diagram
 - AR in log-log plot
 - Phase angle in log-linear plot
 - Via efficient numerical technique (fast Fourier transform, FFT), the output can be calculated for any type of input.
- Frequency response representation of a system dynamics is very convenient for designing a feedback controller and analyzing a closed-loop system.
 - Bode stability
 - Gain margin (GM) and phase margin (PM)

Example

- If a feed is pumped by a peristaltic pump to a CSTR, will the fluctuation of the feed flow appear in the output?

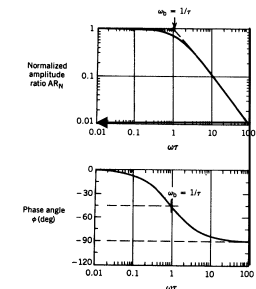


$$V \frac{dc_A}{dt} = q_i c_{Ai} - q c_A \quad (q \approx \text{constant})$$

$$\frac{C_A(s)}{q_i(s)} = \frac{C_{Ai}}{Vs + q} = \frac{C_{Ai}/q}{(V/q)s + 1}$$

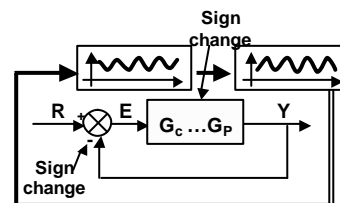
- $V=50\text{cm}^3$, $q=90\text{cm}^3/\text{min}$ (so is the average of q_i)
 - Process time constant = 0.555 min.
- The rpm of the peristaltic pump is 60 rpm.
 - Input frequency = 180 rad/min (3 blades)
- The AR = 0.01 ($\omega t = 100$)

If the magnitude of fluctuation of q_i is 5% of nominal flow rate, the fluctuation in the output concentration will be about 0.05% which is almost unnoticeable.



Critical frequency

- As controller gain changes, the amplitude ratio (AR) and the phase angle (PA) change.
- The frequency where the PA reaches -180° is called critical frequency (ω_c).
- The component of output at the critical frequency will have the exactly same phase as the signal goes through the loop due to comparator (-180°) and phase shift of the process (-180°).
- For the open-loop gain at the critical frequency, $K_{OL}(\omega_c) = 1$
 - No change in magnitude
 - Continuous cycling
- For $K_{OL}(\omega_c) > 1$
 - Getting bigger in magnitude
 - Unstable
- For $K_{OL}(\omega_c) < 1$
 - Getting smaller in magnitude
 - Stable



OBTAINING FREQUENCY RESPONSE

- From the transfer function, replace s with $j\omega$

$$\begin{array}{ccc} G(s) & \xrightarrow{s=j\omega} & G(j\omega) \\ \text{Transfer function} & & \text{Frequency response} \end{array}$$

- For a pole, $s = a + j\omega$, the response mode is $e^{(a+j\omega)t}$.
- If the modes are not unstable ($a \leq 0$) and enough time elapses, the survived modes becomes $e^{j\omega t}$. (ultimate response)

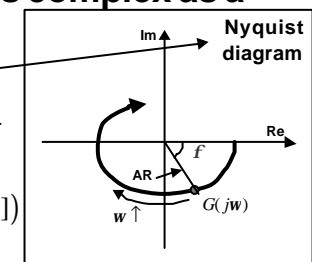
- The frequency response, $G(j\omega)$ is complex as a function of frequency.

$$G(j\omega) = \text{Re}[G(j\omega)] + j \text{Im}[G(j\omega)]$$

$$AR = |G(j\omega)| = \sqrt{\text{Re}[G(j\omega)]^2 + \text{Im}[G(j\omega)]^2}$$

$$f = \angle G(j\omega) = \tan^{-1}(\text{Im}[G(j\omega)]/\text{Re}[G(j\omega)])$$

→ Bode plot



Getting ultimate response

For a sinusoidal forcing function $Y(s) = G(s) \frac{Aw}{s^2 + w^2}$

Assume $G(s)$ has stable poles b_i

$$Y(s) = G(s) \frac{Aw}{s^2 + w^2} = \frac{a_1}{s + b_1} + \dots + \frac{a_n}{s + b_n} + \frac{Cs + Dw}{s^2 + w^2}$$

Decayed out at large t

$$G(jw)Aw = Cjw + Dw \Rightarrow G(jw) = \frac{D}{A} + j\frac{C}{A} = R + jI$$

$$C = IA, D = RA \Rightarrow y_{ul} = A \cos wt + R \sin wt = \hat{A} \sin(wt + f)$$

$$\therefore AR = \hat{A} / A = \sqrt{R^2 + I^2} = |G(jw)| \text{ and } f = \tan^{-1}(I/R) = \angle G(jw)$$

- Without calculating transient response, the frequency response can be obtained directly from $G(jw)$.
- Unstable transfer function does not have a frequency response because a sinusoidal input produces an unstable output response.

Process Zero (lead)

$$G(s) = t_a s + 1$$

$$G(jw) = 1 + jw t_a$$

$$AR_N = |G(jw)| = \sqrt{1 + w^2 t_a^2}$$

$$f = \angle G(jw) = \tan^{-1}(w t_a)$$

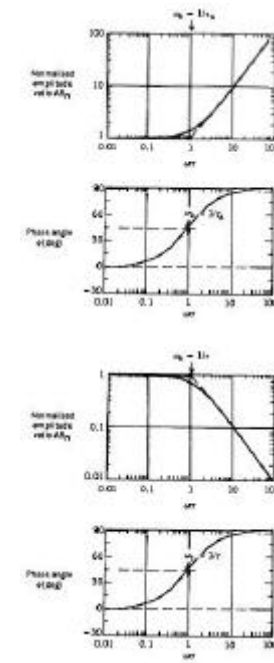
Unstable pole

$$G(s) = \frac{1}{(-t_s s + 1)}$$

$$G(jw) = \frac{1}{1 - jtw} = \frac{1}{1 + t^2 w^2} (1 + jtw)$$

$$AR = |G(jw)| = \frac{1}{\sqrt{1 + w^2 t^2}}$$

$$f = \angle G(jw) = \tan^{-1} \frac{\text{Im}(G(jw))}{\text{Re}(G(jw))} = \tan^{-1} wt$$



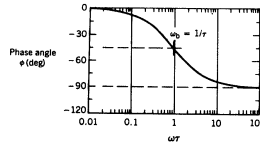
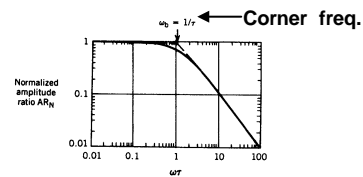
First-order process

$$G(s) = \frac{K}{(ts + 1)}$$

$$G(jw) = \frac{K}{(1 + jw t)} = \frac{K}{(1 + w^2 t^2)} (1 - jw t)$$

$$AR_N = |G(jw)| = \frac{1}{\sqrt{1 + w^2 t^2}}$$

$$f = \angle G(jw) = -\tan^{-1}(wt)$$



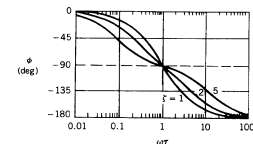
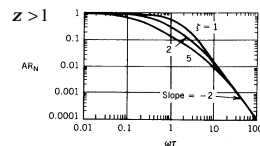
Second-order process

$$G(s) = \frac{K}{(t^2 s^2 + 2zts + 1)}$$

$$G(jw) = \frac{K}{(1 - t^2 w^2) + 2jzwt}$$

$$AR = |G(jw)| = \frac{K}{\sqrt{(1 - w^2 t^2)^2 + (2zwt)^2}}$$

$$f = \angle G(jw) = \tan^{-1} \frac{\text{Im}(G(jw))}{\text{Re}(G(jw))} = -\tan^{-1} \frac{2zwt}{1 - w^2 t^2}$$



Integrating process

$$G(s) = \frac{1}{As} \quad G(jw) = \frac{1}{jAw} = -\frac{1}{Aw} j$$

$$AR_N = |G(jw)| = \frac{1}{Aw}$$

$$f = \angle G(jw) = \tan^{-1}\left(-\frac{1}{0 \cdot w}\right) = -\frac{p}{2}$$

Differentiator

$$G(s) = As \quad G(jw) = jAw$$

$$AR_N = |G(jw)| = Aw$$

$$f = \angle G(jw) = \tan^{-1}\left(\frac{1}{0 \cdot w}\right) = \frac{p}{2}$$

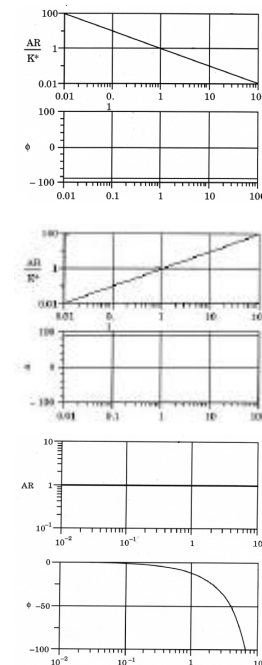
Pure delay process

$$G(s) = e^{-qs}$$

$$G(jw) = e^{-jqw} = \cos qw - j \sin qw$$

$$AR = |G(jw)| = 1$$

$$f = \angle G(jw) = -\tan^{-1} \tan qw = -qw$$



SKETCHING BODE PLOT

$$G(s) = \frac{G_a(s)G_b(s)G_c(s)\cdots}{G_1(s)G_2(s)G_3(s)\cdots} \quad G(j\omega) = \frac{G_a(j\omega)G_b(j\omega)G_c(j\omega)\cdots}{G_1(j\omega)G_2(j\omega)G_3(j\omega)\cdots}$$

$$|G(j\omega)| = \frac{|G_a(j\omega)||G_b(j\omega)||G_c(j\omega)|\cdots}{|G_1(j\omega)||G_2(j\omega)||G_3(j\omega)|\cdots}$$

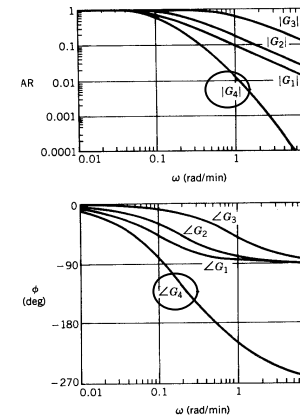
$$\angle G(j\omega) = \angle G_a(j\omega) + \angle G_b(j\omega) + \angle G_c(j\omega) + \cdots \\ - \angle G_1(j\omega) - \angle G_2(j\omega) - \angle G_3(j\omega) - \cdots$$

Boddiagram

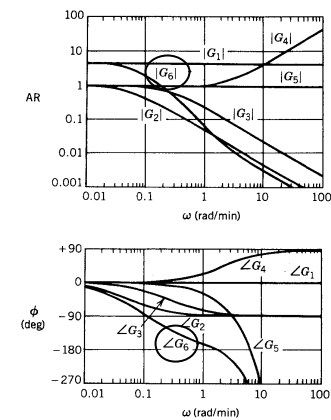
- AR vs. frequency in log-log plot
- PA vs. frequency in semi-log plot
- Useful for
 - Analysis of the response characteristics
 - Stability of the closed-loop system only for open-loop stable systems with phase angle curves exhibit a single critical frequency.

Examples

$$1. G(s) = \frac{K}{(10s+1)(5s+1)(s+1)} \quad \begin{matrix} G_1 & G_2 & G_3 \end{matrix}$$



$$2. G(s) = \frac{G_1 G_5 G_4}{G_2 G_3} = \frac{5(0.5s+1)e^{-0.5s}}{(20s+1)(4s+1)}$$



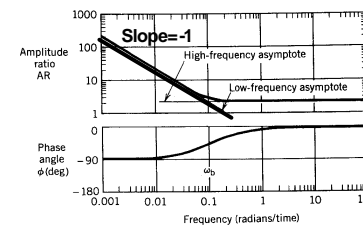
Amplitude Ratio on log-log plot

- Start from steady-state gain at $w=0$. If G_{OL} includes either integrator or differentiator it starts at ∞ or 0 .
- Each first-order lag (lead) adds to the slope -1 ($+1$) starting at the corner frequency.
- Each integrator (differentiator) adds to the slope -1 ($+1$) starting at zero frequency.
- A delay does not contribute to the AR plot.

Phase angle on semi-log plot

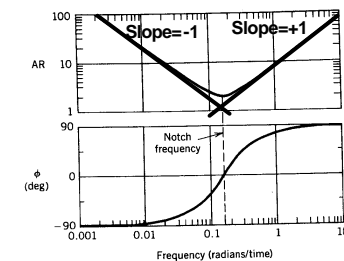
- Start from 0° or -180° at $w=0$ depending on the sign of steady-state gain.
- Each first-order lag (lead) adds 0° to phase angle at $w=0$, adds -90° ($+90^\circ$) to phase angle at $w=\infty$, and adds -45° ($+45^\circ$) to phase angle at corner frequency.
- Each integrator (differentiator) adds -90° ($+90^\circ$) to the phase angle for all frequency.
- A delay adds $-qw$ to phase angle depending on the frequency.

$$3. \text{PI: } G(s) = K_C \left(1 + \frac{1}{t_I s} \right)$$



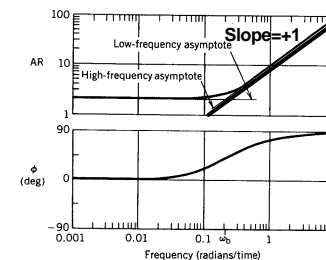
$$w_b = 1/t_I \text{ at } f = -45^\circ$$

$$5. \text{PID: } G(s) = K_C \left(1 + \frac{1}{t_I s} + t_D s \right)$$



$$w_{\text{Notch}} = 1/\sqrt{t_I t_D} \text{ at } f = 0^\circ$$

$$4. \text{PD: } G(s) = K_C (1 + t_D s)$$



$$w_b = 1/t_D \text{ at } f = -45^\circ$$

NYQUIST DIAGRAM

- **Alternative representation of frequency response**
- **Polar plot of $G(j\omega)$ (ω is implicit)**

$$G(j\omega) = \text{Re}[G(j\omega)] + j \text{Im}[G(j\omega)]$$

- **Compact (one plot)**
- **Wider applicability of stability analysis than Bode plot**
- **High frequency characteristics will be shrunk near the origin.**
 - **Inverse Nyquist diagram: polar plot of $1/G(j\omega)$**
- **Combination of different transfer function components is not easy as with Nyquist diagram as with Bode plot.**

